

## 1000mA Synchronous Step-Down Converter

### Description

The SC11A10 is high efficiency synchronous, PWM step-down DC/DC converters working under an input voltage range from 2.5V to 5.5V. This feature makes the SC11A10 suitable for single Li-Ion battery-powered applications. 100% duty cycle capability extends battery life in portable devices, while the quiescent current is 120 $\mu$ A with no load, and drops to < 1 $\mu$ A in shutdown

The internal synchronous switch is designed to increase efficiency without an external Schottky diode. The 1.2MHz fixed switching frequency allows the using of tiny, low profile inductors and ceramic capacitors, which minimized overall solution footprint.

The SC11A10 converters are available in the industry standard SOT-23-5 power packages (or upon request).

### Features

- Up to 93% Efficiency
- Current Mode Operation for Excellent Line and Load Transient Response
- Low Quiescent Current: 160 $\mu$ A
- Low Switch on Resistance RDS(ON), Internal Switch: 0.35 $\Omega$
- Output Voltage: 0.6V ~ 5.5V
- No Schottky Diode Required
- 1.0MHz Fixed Frequency Switching
- Short-Circuit Protection
- Shutdown Quiescent Current: < 1 $\mu$ A
- Low Profile SOT-23-5 Package (lead-free packaging is now available)

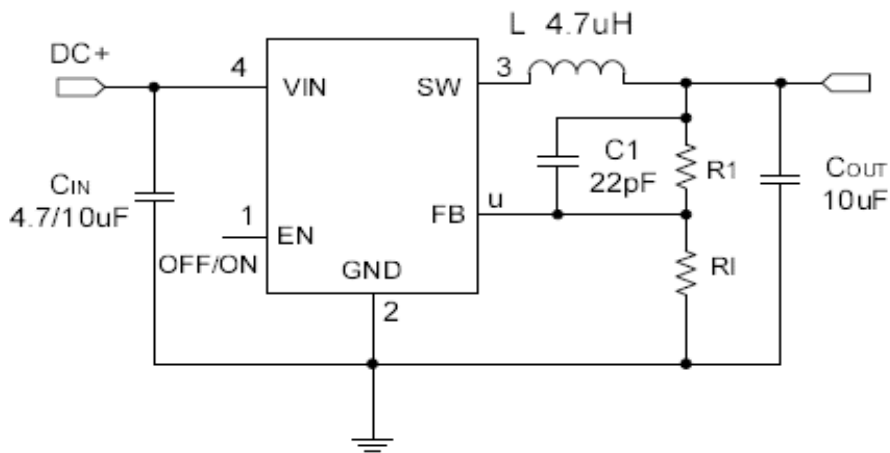
### Applications

- Digital cameras and MP3
- Palmtop computers / PDAs
- Cellular phones
- Wireless handsets and DSL modems
- PC cards
- Portable media players

### Pin Configurations

Package Type	Pin Configurations
SC11A10 SOT-23-5	

### Typical Application Circuit



**Figure 1 : Adjustable Output Voltage**  
 $V_{OUT} = 0.6V \cdot [1 + (R1/R2)]$

## Pin Description

PIN SOT-23-5	NAME	DESCRIPTION
1.	EN	En Control Input. Forcing this pin above 1.5V enables the part. Forcing this pin below 0.3V can shut down the device. In shutdown, all functions are disabled drawing <1mA supply current. Do not leave EN floating.
2.	GND	Ground.
3.	SW	Switch Node Connection to Inductor. This pin connects to the drains of the internal main and synchronous power MOSFET switches.
4.	VIN	Main Supply Pin. It must be closely decoupled to GND, Pin 2, with a 10 $\mu$ F or greater ceramic capacitor.
5.	FB	Feedback Pin. Receive the feedback voltage from an external resistive divider across the output. In the adjustable version, the output voltage is set by a resistive divider according to the following formula: $V_{OUT} = 0.6V \cdot [1 + (R1/R2)]$ .

## Absolute Maximum Ratings

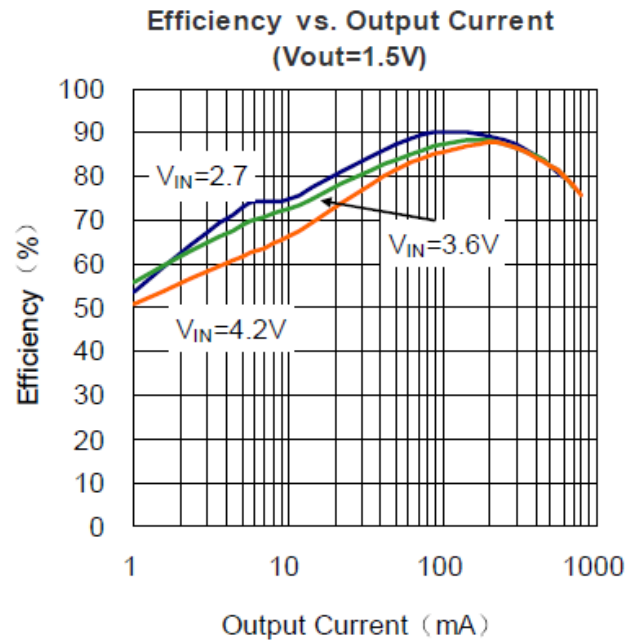
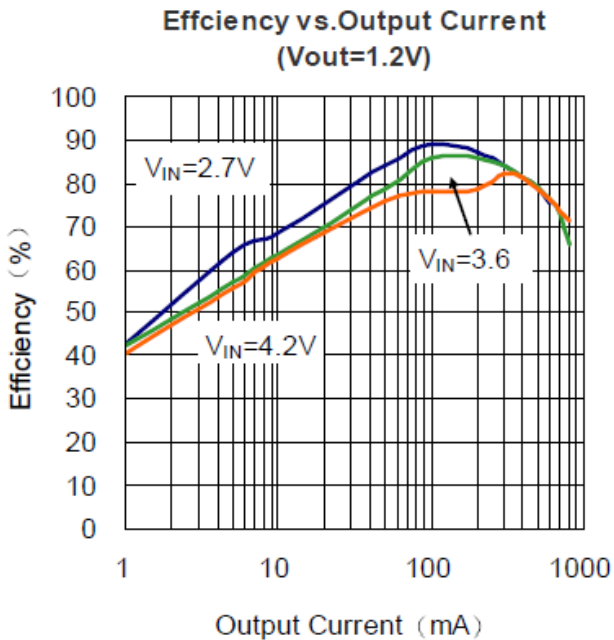
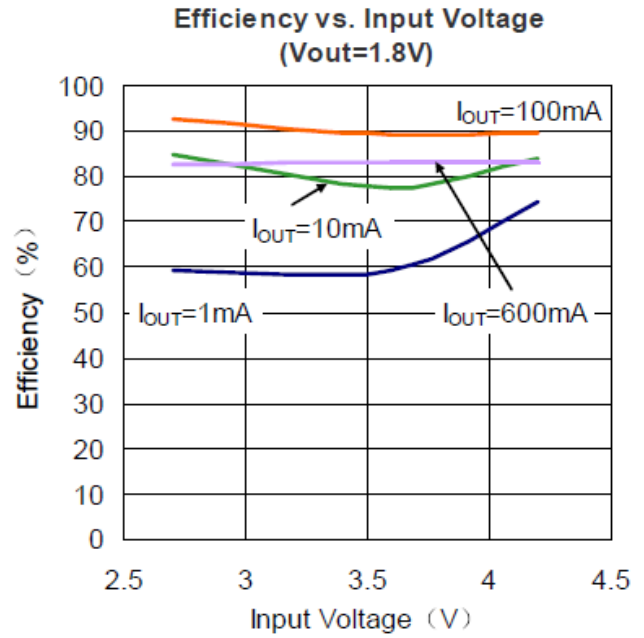
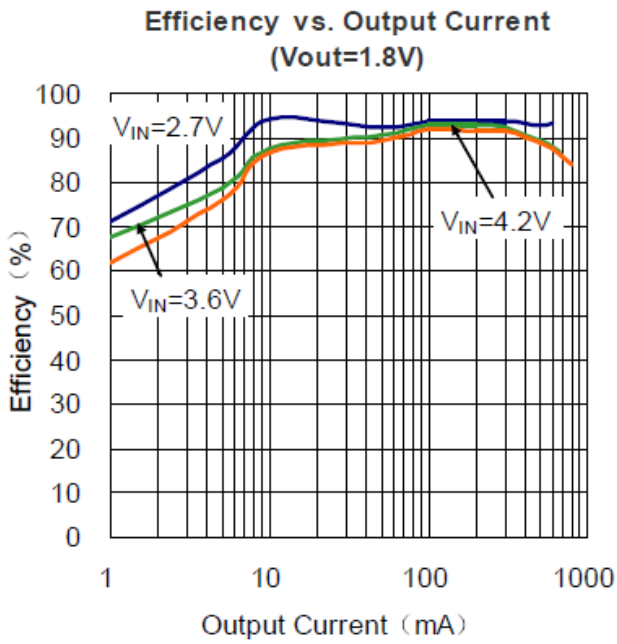
- Input Supply Voltage (VIN), VFB ----- -0.3V to 6V
- VON/OFF, VSW. ----- - 0.3V to VIN + 0.3V
- ISW ----- 1.3A
- Operating Ambient Temperature Range ----- -40°C to 85°C
- Storage Temperature Range ----- -65°C to 125°C
- Lead Temperature (Soldering, 10 sec) ----- 300°C
- Junction Temperature ----- 125°C

## Electrical Characteristics

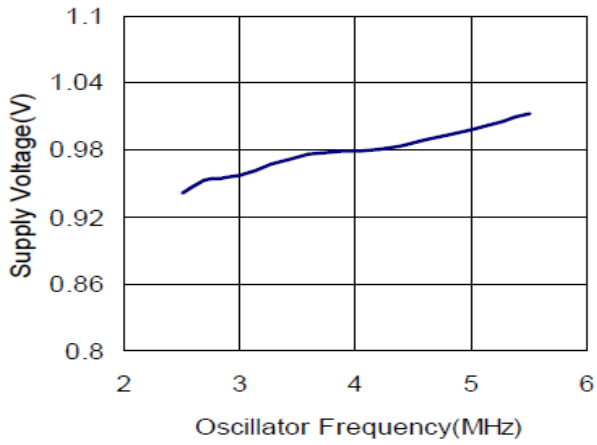
(Operating Conditions:  $T_A=25^{\circ}\text{C}$ ,  $V_{IN}=3.6\text{V}$ )

SYMBOL	PARAMETER	CONDITIONS	SC11A10			UNITS
			MIN	TYP	MAX	
$V_{OUT}$	Output Voltage	$I_{OUT} = 100\text{mA}$ , $R1/R2=2$	1.75	1.8	1.85	V
$V_{IN}$	Input Voltage Range		2.5		5.5	V
$V_{FB}$	Regulated Voltage	$T_A = 25^{\circ}\text{C}$	0.588	0.6	0.612	V
$I_{FB}$	Feedback Current				$\pm 30\%$	nA
$\Delta V_{FB}$	VREF	$V_{IN}=2.5\text{V}\sim 5.5\text{V}$		0.03	0.4	%/V
$F_{OSC}$	Oscillator Frequency	$V_{FB} = 0.6\text{V}$ or $V_{OUT} = 100\%$	1.0	1.2	1.4	MHz
$I_Q$	Quiescent Current	$V_{FB} = 0.5\text{V}$ or $V_{OUT} = 90\%$ , $I_{LOAD} = 0\text{A}$		120		$\mu\text{A}$
$I_S$	Shutdown Current	$V_{EN} = 0\text{V}$ , $V_{IN} = 4.2\text{V}$		0.1	1	$\mu\text{A}$
$I_{PK}$	Peak Inductor Current	$V_{IN} = 3\text{V}$ , $V_{FB} = 0.5\text{V}$ or $V_{OUT} = 90\%$ , Duty Cycle < 35%	0.95	1		A
$R_{PFET}$	$R_{DS(ON)}$ of P-Channel FET	$I_{SW} = 100\text{mA}$		0.3		$\Omega$
$R_{NFET}$	$R_{DS(ON)}$ of N-Channel FET	$I_{SW} = -100\text{mA}$		0.39		$\Omega$
$EFFI$	Efficiency	When connected to ext. components $V_{IN}=V_{EN}=3.6\text{V}$ , $I_{OUT}=100\text{mA}$		93		%
$\Delta V_{OUT}$	$V_{OUT}$ line Regulation	$V_{IN}=2.5\text{V}\sim 5.5\text{V}$		0.03	0.3	%/V
$V_{LOADREG}$	$V_{OUT}$ load Regulation			0.33		%

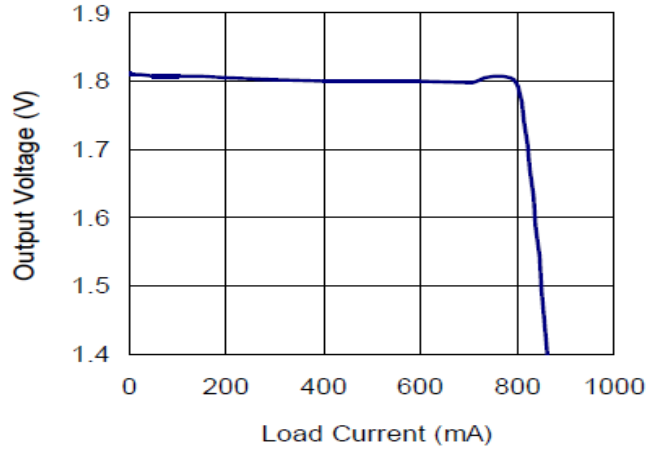
## Typical Performance Characteristics



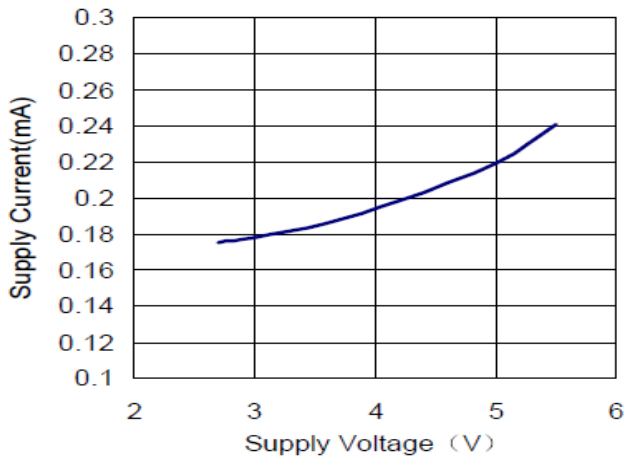
**Oscillator Frequency vs. Supply Voltage (Vout=1.8V Io=100mA)**



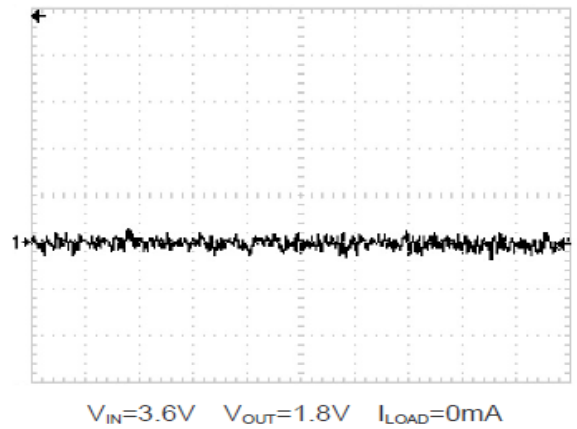
**Output Voltage vs. Load Current (Vin=3.6V)**



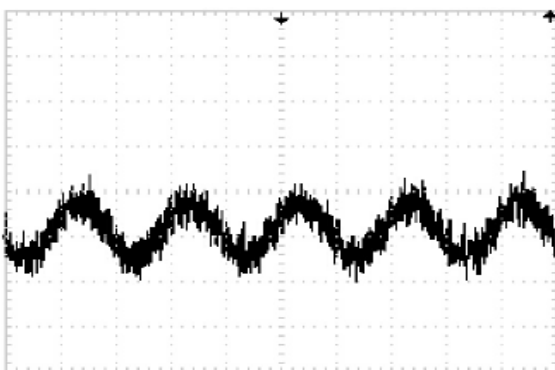
**Supply Current vs Supply Voltage (Vout=1.8V Io=0A)**



**Output Noise (100mV/DIV 2ms/DIV AC COUPLED)**

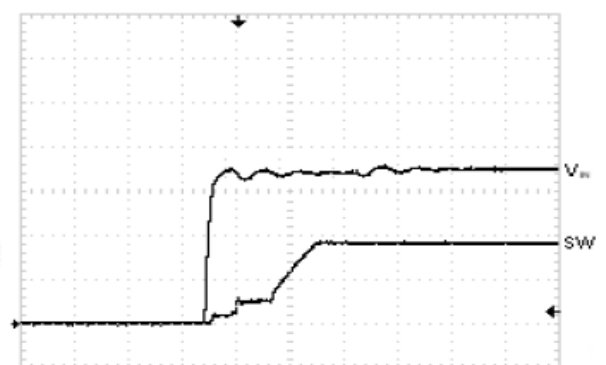


**Output Noise (20mV/DIV 10ms/DIV AC COUPLED)**



V<sub>IN</sub>=3.6V V<sub>OUT</sub>=1.8V I<sub>LOAD</sub>=200mA

**Start-up from Shutdown Input and Output Noise (1V/DIV 100ns/DIV)**



## Application Information

The basic SC11A10 application circuit is shown in Typical Application Circuit. External component selection is determined by the maximum load current and begins with the selection of the inductor value and operating frequency followed by C<sub>IN</sub> and C<sub>OUT</sub>.

### Inductor Selection

For most applications, the value of the inductor will fall in the range of 1mH to 4.7mH. Its value is chosen based on the desired ripple current. Large value inductors lower ripple current and small value inductors result in higher ripple currents. Higher V<sub>IN</sub> or V<sub>OUT</sub> also increases the ripple current as shown in equation 1. A reasonable starting point for setting ripple current is ΔI<sub>L</sub>=360mA (40% of 900mA).

$$\Delta I_L = \frac{1}{(f)(L)} V_{OUT} \left( 1 - \frac{V_{OUT}}{V_{IN}} \right)$$

The DC current rating of the inductor should be at least equal to the maximum load current plus half the ripple current to prevent core saturation. Thus, a 1.08A rated inductor should be enough for most applications (900mA + 180mA). For better efficiency, choose a low DC-resistance inductor.

Different core materials and shapes will change the size/current and price/current relationship of an inductor. Toroid or shielded pot cores in ferrite or perm alloy materials are small and don't radiate much energy, but generally cost more than powdered iron core inductors with similar electrical characteristics. The choice of which style inductor to use often depends more on the price vs. size requirements and any radiated field/EMI requirements than on what the SC11A10 requires to operate.

### Output and Input Capacitor Selection

In continuous mode, the source current of the top MOSFET is a square wave of duty cycle V<sub>OUT</sub>/V<sub>IN</sub>. To prevent large voltage transients, a low ESR input capacitor sized for the maximum RMS current must be used. The maximum RMS capacitor current is given by:

$$C_{IN} \text{ required } I_{RMS} \approx I_{OMAX} \frac{[V_{OUT}(V_{IN} - V_{OUT})]^{1/2}}{V_{IN}}$$

This formula has a maximum at V<sub>IN</sub>= 2V<sub>OUT</sub>, where I<sub>RMS</sub>=I<sub>OUT</sub>/2. This simple worst-case condition is commonly used for design because even significant deviations do not offer much relief. Note that the capacitor manufacturer's ripple current ratings are often based on 2000 hours of life. This makes it advisable to further derate the capacitor, or choose a capacitor rated at a higher temperature than required. Always consult the manufacturer if there is any question. The selection of C<sub>OUT</sub> is driven by the required effective series resistance (ESR). Typically, once the ESR requirement for C<sub>OUT</sub> has been met, the RMS current rating generally far exceeds the I<sub>ripple(P-P)</sub> requirement. The output ripple ΔV<sub>OUT</sub> is determined by:

$$\Delta V_{OUT} \approx \Delta I_L \left( ESR + \frac{1}{8fC_{OUT}} \right)$$

Where f=operating frequency, C<sub>OUT</sub>=output capacitance and ΔI<sub>L</sub>=ripple current in the inductor. For a fixed output voltage, the output ripple is highest at maximum input voltage since ΔI<sub>L</sub> increases with input voltage.

Aluminum electrolytic and dry tantalum capacitors are both available in surface mount configurations. In the case of tantalum, it is critical that the capacitors are surge tested for use in switching power supplies. An excellent choice is the AVX TPS series of surface mount tantalum. These are specially constructed and

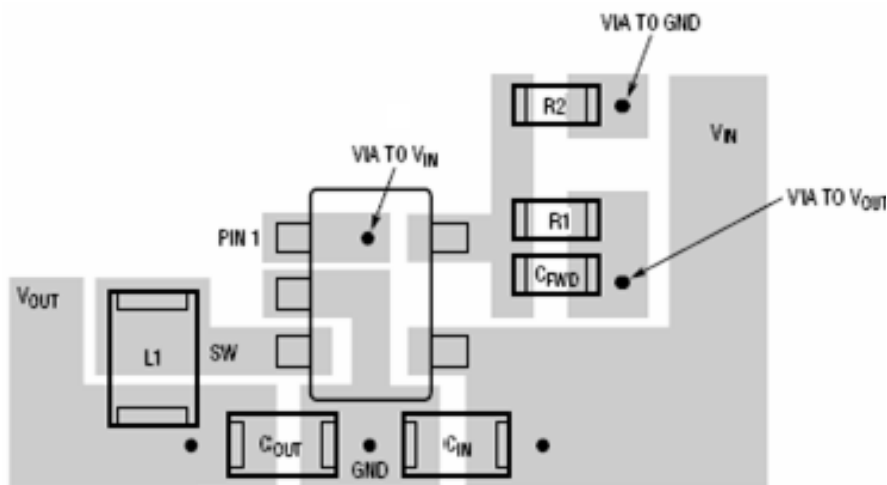
tested for low ESR so they give the lowest ESR for a given volume. Other capacitor types include Sanyo POSCAP, Kemet T510 and T495 series, and Sprague 593D and 595D series. Consult the manufacturer for other specific recommendations.

### PCB Layout Guidelines

When laying out the printed circuit board, the following checklist should be used to ensure proper operation of the SC11A10. These items are also illustrated graphically in Figure 1. Check the following in your layout:

- The power traces, consisting of the GND trace, the SW trace and the VIN trace should be kept short, direct and wide.
- Does the VFB pin connect directly to the feedback resistors? The resistive divider R1/R2 must be connected between the (+) plate of C<sub>OUT</sub> and ground.
- Does the (+) plate of C<sub>IN</sub> connects to V<sub>IN</sub> as closely as possible? This capacitor provides the AC current to the internal power MOSFETs.
- Keep the switching node, SW, away from the sensitive VFB node.
- Keep the (-) plates of C<sub>IN</sub> and C<sub>OUT</sub> as close as possible.

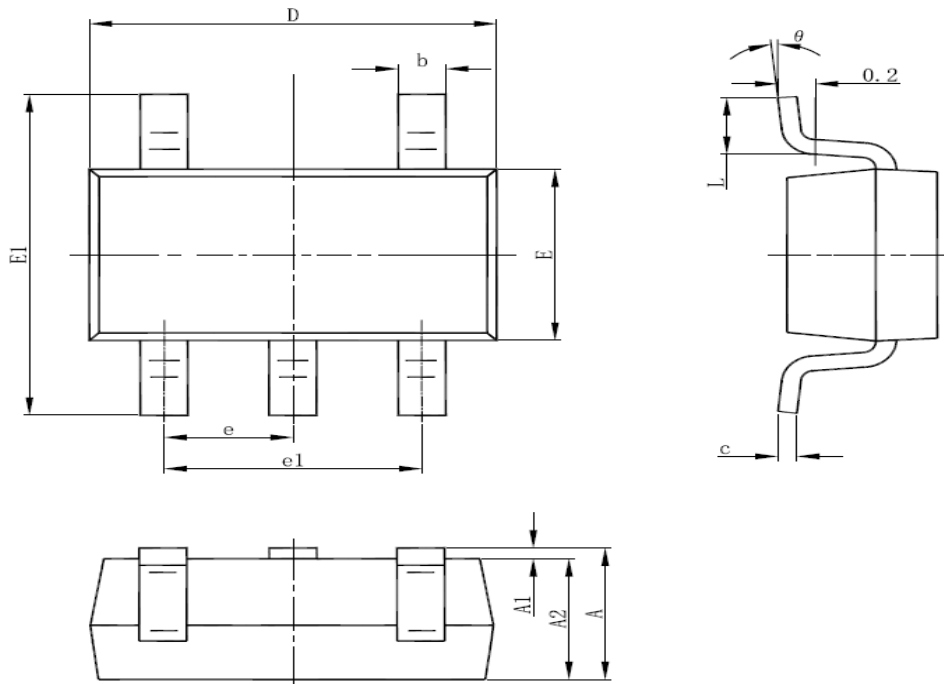
### PCB Layout Recommend





## Packaging Information

### SPT-23-5 Package Outline Dimension



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	1.050	1.250	0.041	0.049
A1	0.000	0.100	0.000	0.004
A2	1.050	1.150	0.041	0.045
b	0.300	0.500	0.012	0.020
c	0.100	0.200	0.004	0.008
D	2.820	3.020	0.111	0.119
E	1.500	1.700	0.059	0.067
E1	2.650	2.950	0.104	0.116
e	0.950(BSC)		0.037(BSC)	
e1	1.800	2.000	0.071	0.079
L	0.300	0.600	0.012	0.024
θ	0°	8°	0°	8°